



Toward the Optimal Bit Aspect Ratio in Magnetic Recording

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with contributions from
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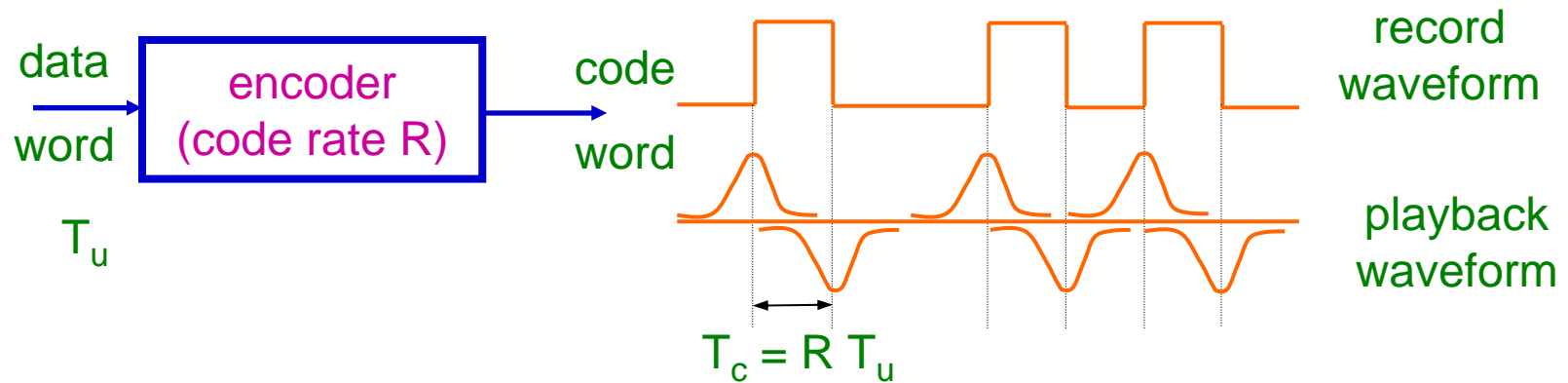


Outline

- ⌘ Background
- ⌘ Channel Model
- ⌘ Approach for Shannon Codes
- ⌘ Optimal Code Rates for Shannon Codes on the Lorentzian Channel
- ⌘ Approach for LDPC Codes
- ⌘ Optimal Code Rates for LDPC Codes on the Lorentzian Channel
- ⌘ On the Optimal Bit Aspect Ratio
- ⌘ Concluding Remarks

Background

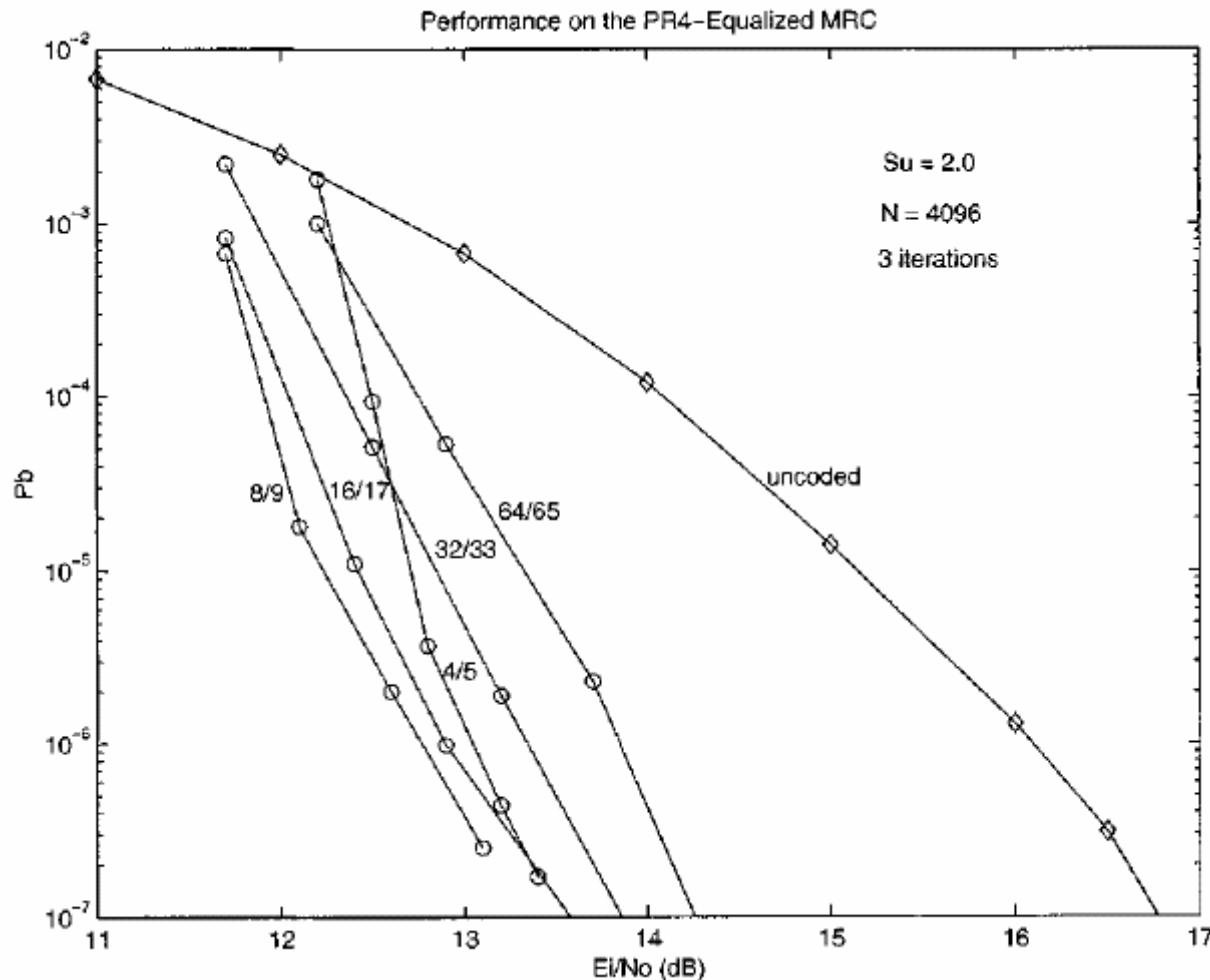
⌘ Coding on a magnetic recording channel: Lorentzian model



- ⌘ due to ISI, the code rate loss is R^2 -- on the AWGN channel it is R
- ⌘ on the AWGN channel, performance improves with decreasing code rate; on ISI channels such as the Lorentzian, it does not

Background (cont'd)

⌘ In [Ryan, Trans. Magn., Nov. 2000] we examined optimal code rates empirically for specific parallel and serial turbo codes



Performance of various PCCC's on the PR4-equalized Lorentzian channel with user density $S_u = 2.0$.

$$S_u = PW_{50}/T_u$$

Channel Model

- ⌘ Lorentzian model (in AWGN)

$$r(t) = \sum_k \frac{1}{2} a_k s(t - kT_c) + w(t)$$

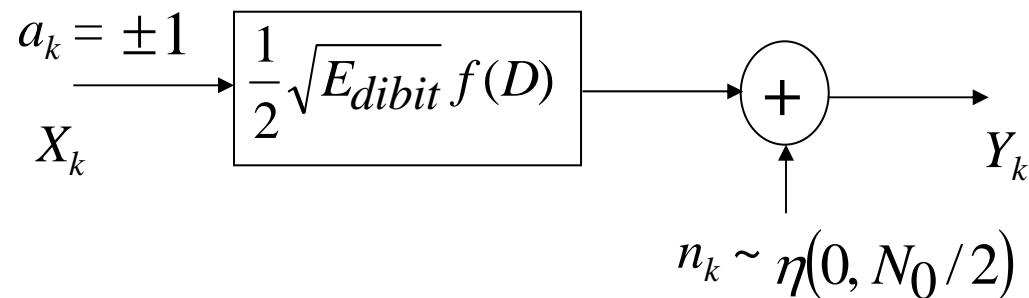
where $s(t) = h(t) - h(t - T_c)$ is the dibit is AWGN with spectral density $N_0/2$ and $h(t)$ is the Lorentzian pulse

$$h(t) = \sqrt{\frac{4E_i}{\pi \text{pw}_{50}}} \frac{1}{1 + (2t / \text{pw}_{50})^2}$$

- ⌘ E_i = the energy per isolated Lorentzian pulse $h(t)$ and pw_{50} is its width measured at half its height

Channel Model (cont'd)

- ⌘ applying a whitened matched filter to $r(t)$ leads to the discrete-time equivalent model depicted below

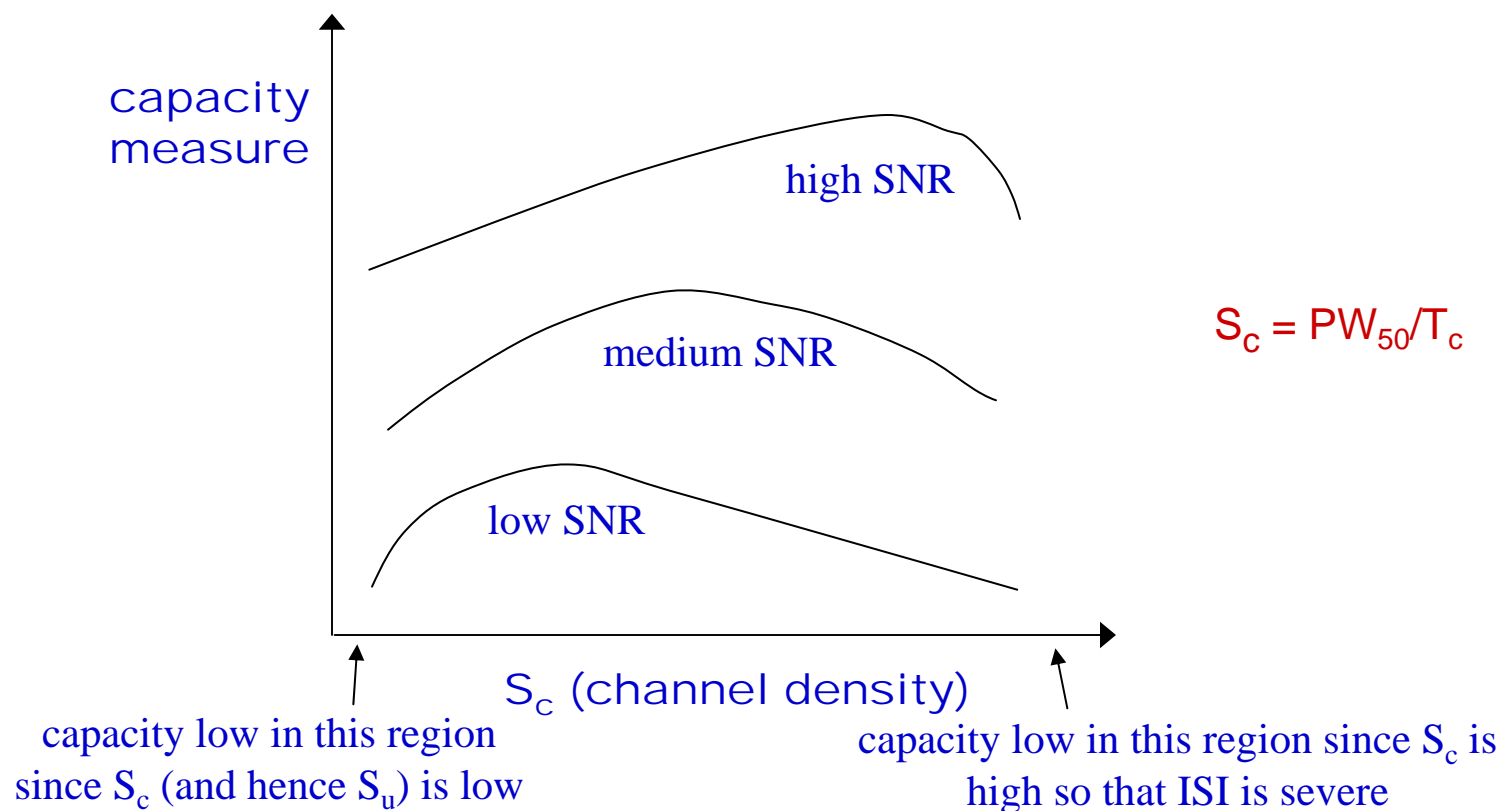


- ⌘ where

- ⊗ E_{dibit} is the energy in $s(t)$,
- ⊗ $f(D)$ is the minimum phase factor in the T_c -sampled auto-correlation function of $s(t)$, $R_s(D)$
- ⊗ $\sum_k f_k^2 = 1$

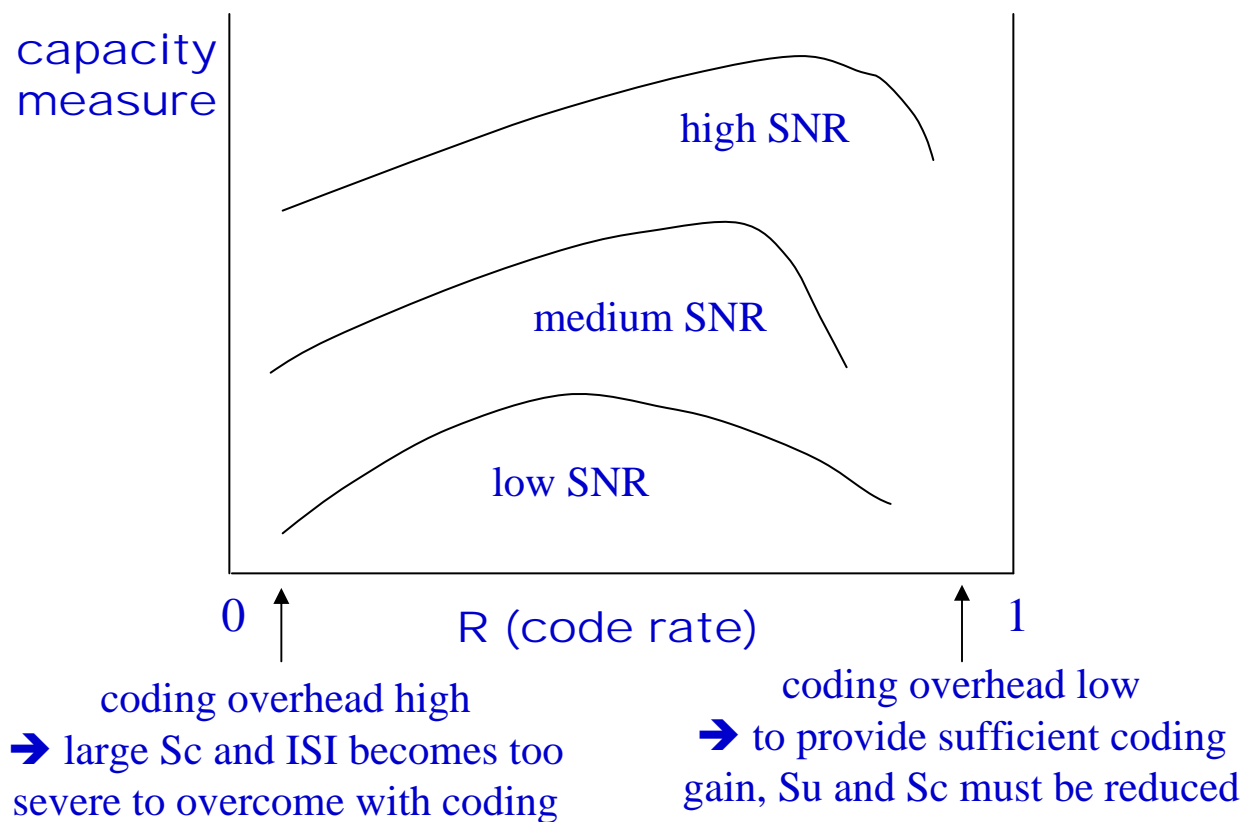
Approach for Shannon Codes

- ⌘ Our goal is to determine optimal code rates for this channel for both Shannon codes and LDPC codes.



Approach for Shannon Codes (cont'd)

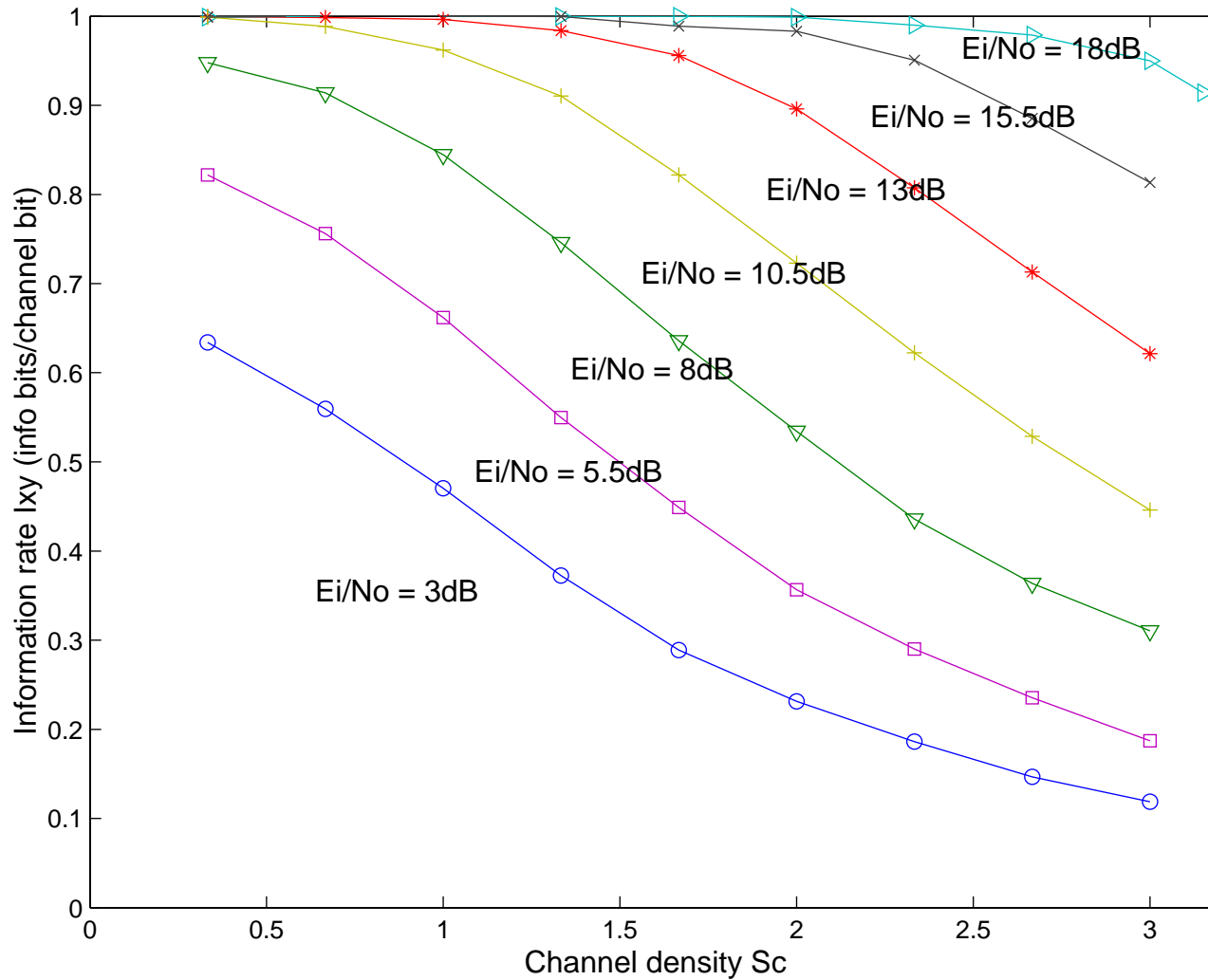
⌘ possibly better is data such as that in the figure below



Approach for Shannon Codes (cont'd)

- ⌘ can now use the result of Arnold-Loeliger (ICC'01) (also, Pfister-Siegel, GC'01) to compute the achievable information rate of the binary-input ISI channel $\frac{1}{2}\sqrt{E_{dibit}} f(D)$ assuming iid inputs
- ⌘ Note by computing the information rate for $\frac{1}{2}\sqrt{E_{dibit}} f(D)$, we do not assume PR equalization. Rather, optimal (ML) detection is assumed.
- ⌘ Note also that we use E_i / N_0 as our SNR measure

Results for Shannon Codes



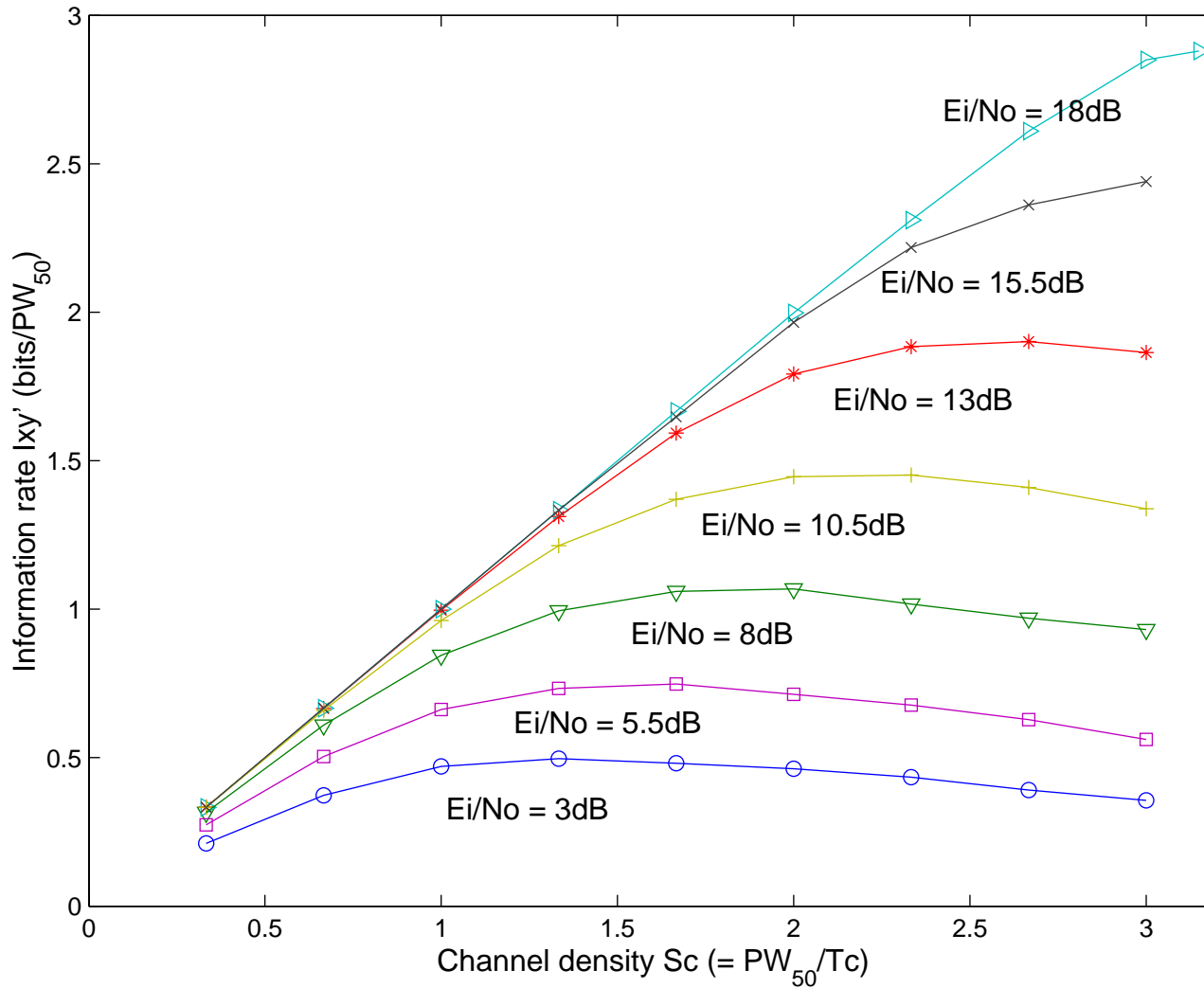
Information rate of Lorentzian channel versus channel density S_c .

Results for Shannon Codes (cont'd)

- ⌘ Note I_{xy} is in units of *information bits/channel bit*
- ⌘ we would like a capacity measure relative to a physical parameter of the channel, such as *info bits/inch* (along a track)
- ⌘ *info bits/pw₅₀* is particularly convenient:
 - ☑ note $S_c = pw_{50}/T_c$ may be regarded as *channel bits/pw₅₀*
 - ☑ (Example: $S_c = 3 \rightarrow 3$ *channel bits/pw₅₀*)
- ⌘ define a new information rate

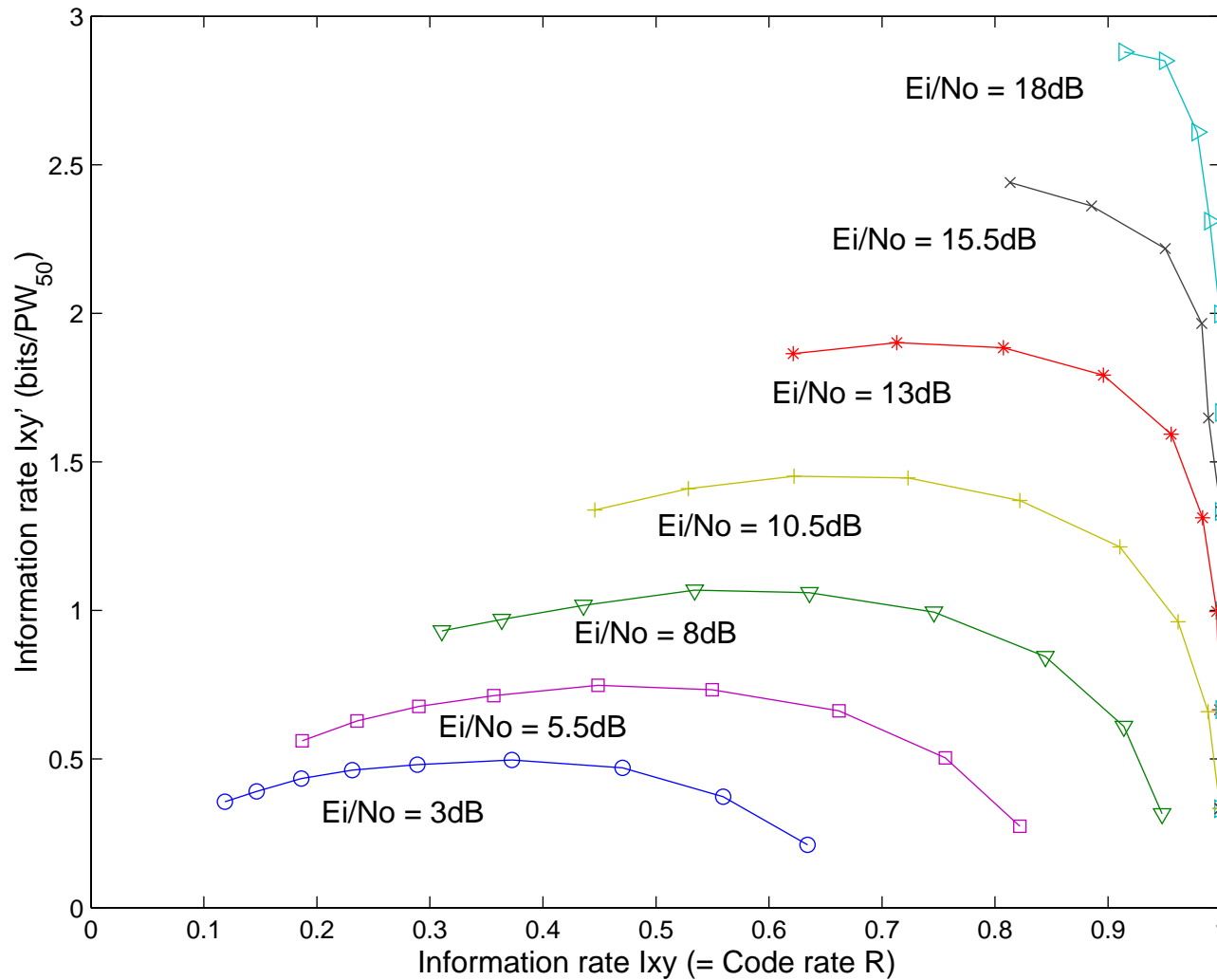
$$I'_{xy} \left(\frac{\text{info bits}}{pw_{50}} \right) = I_{xy} \left(\frac{\text{info bits}}{\text{channel bit}} \right) \cdot S_c \left(\frac{\text{channel bits}}{pw_{50}} \right)$$

Results for Shannon Codes (cont'd)



Normalized
Information rate of
Lorentzian channel
versus channel
density S_c .

Results for Shannon Codes (cont'd)



Information rate of Lorentzian channel versus code rate R.

Results for Shannon Codes (cont'd)

EXAMINATION OF THE DEVIATION OF $I(X;Y)$ WITH κ_{max}

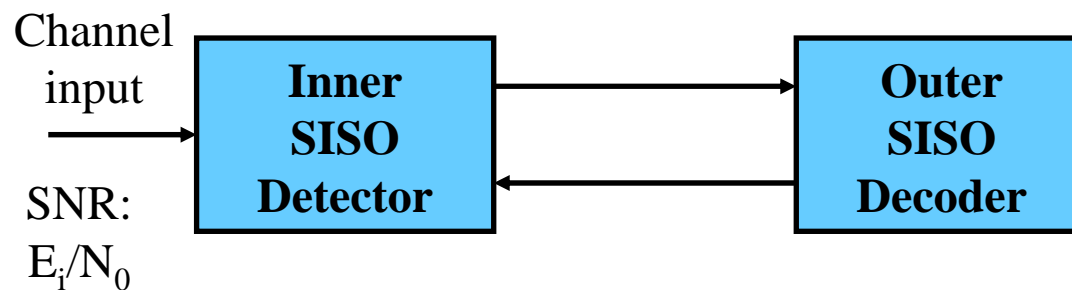
S_c	κ_{max}	L	E_i/N_0 (dB)	$I(X;Y)$	$\% \Delta_{max}$
3	18	15	3	0.1368	3.04
	19			0.1328	
	23			0.1353	
	19		8	0.3057	4.75
	25			0.3025	
	30			0.3169	
	18		13	0.6264	1.53
	25			0.6211	
	30			0.6306	

Examination of deviation of $I(X;Y)$ with $R_s(D)$ truncation parameter κ_{max}

Approach for LDPC Codes

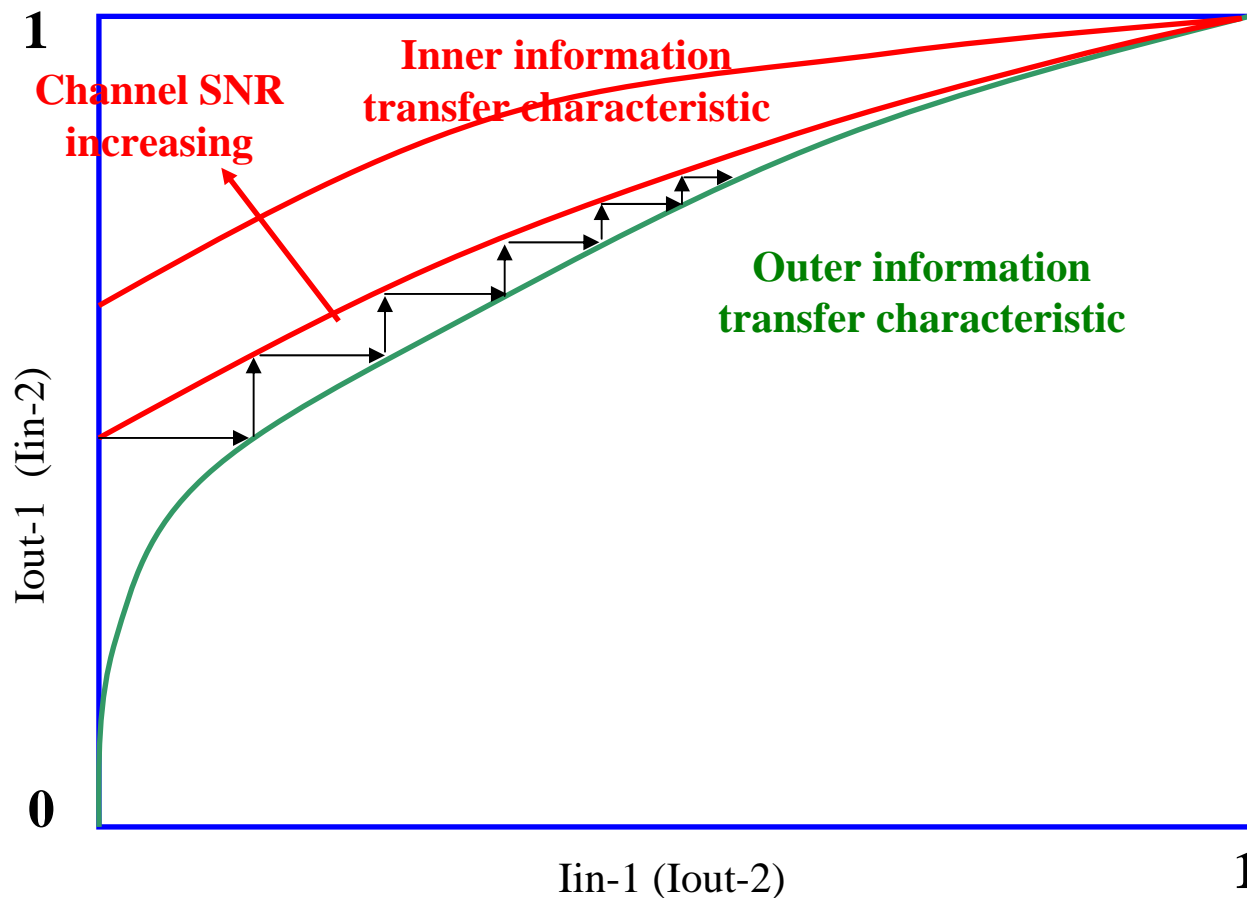
⌘ Extrinsic information transfer (EXIT) chart

- ☒ provides a simple way of determining the capacity limit (or decoding threshold) for a specific coding scheme.
- ☒ describes the flow of extrinsic information through SISO processors (detectors/decoders) operating cooperatively and iteratively.



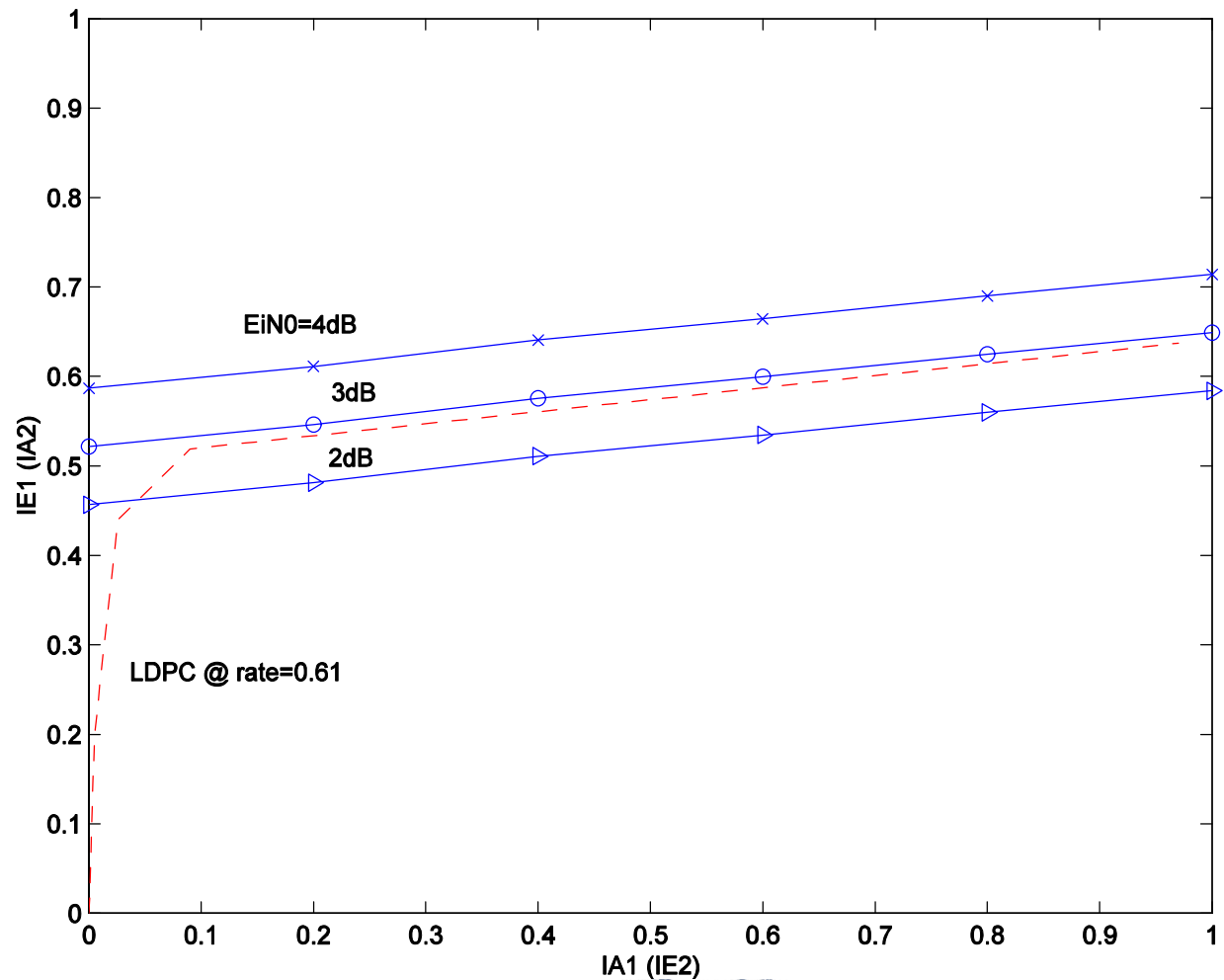
Approach for LDPC Codes (cont'd)

⌘ possibly better is data such as that in the figure below



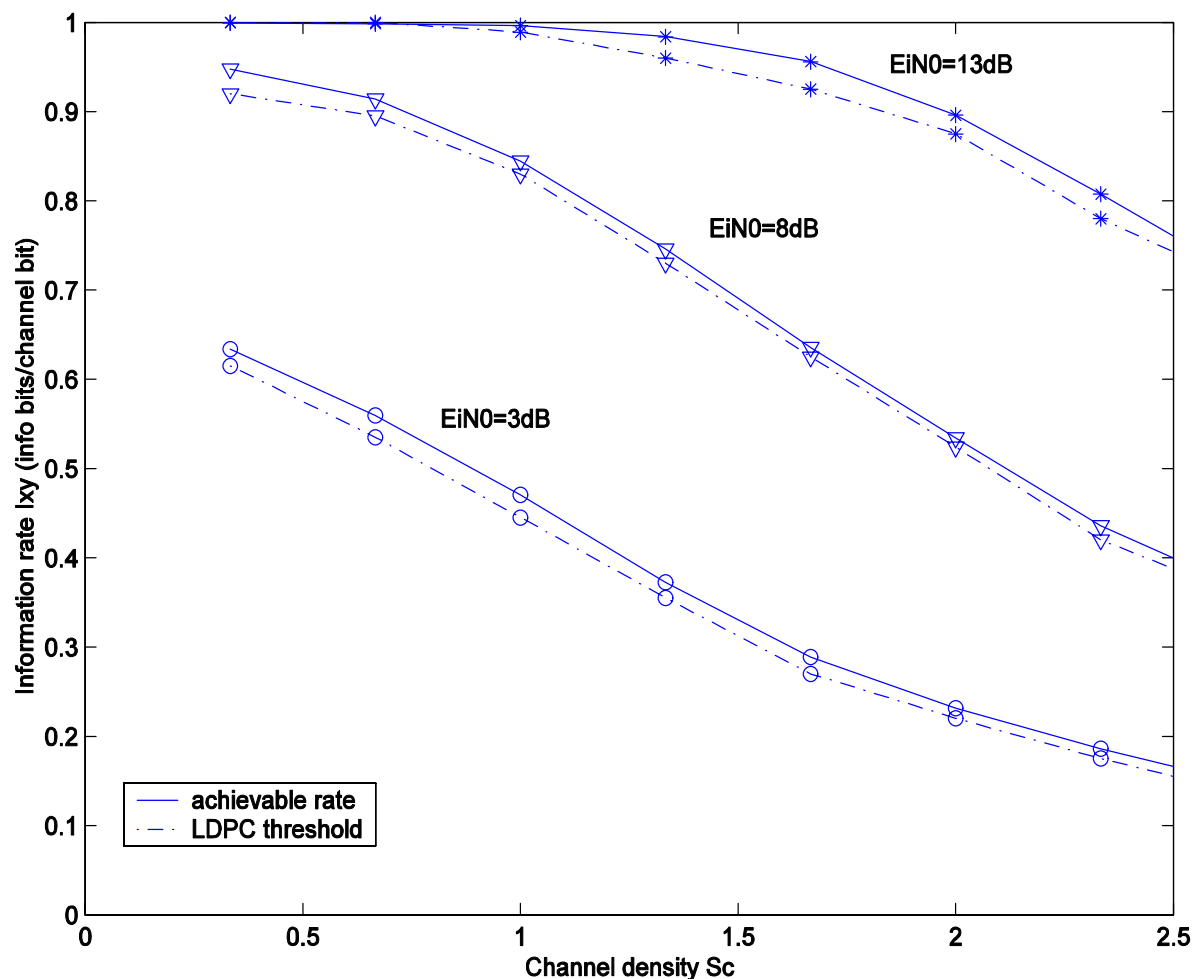
Approach for LDPC Codes (cont'd)

EXIT chart
for channel
density
 $S_c=1/3$ and
LDPC code
rate 0.61



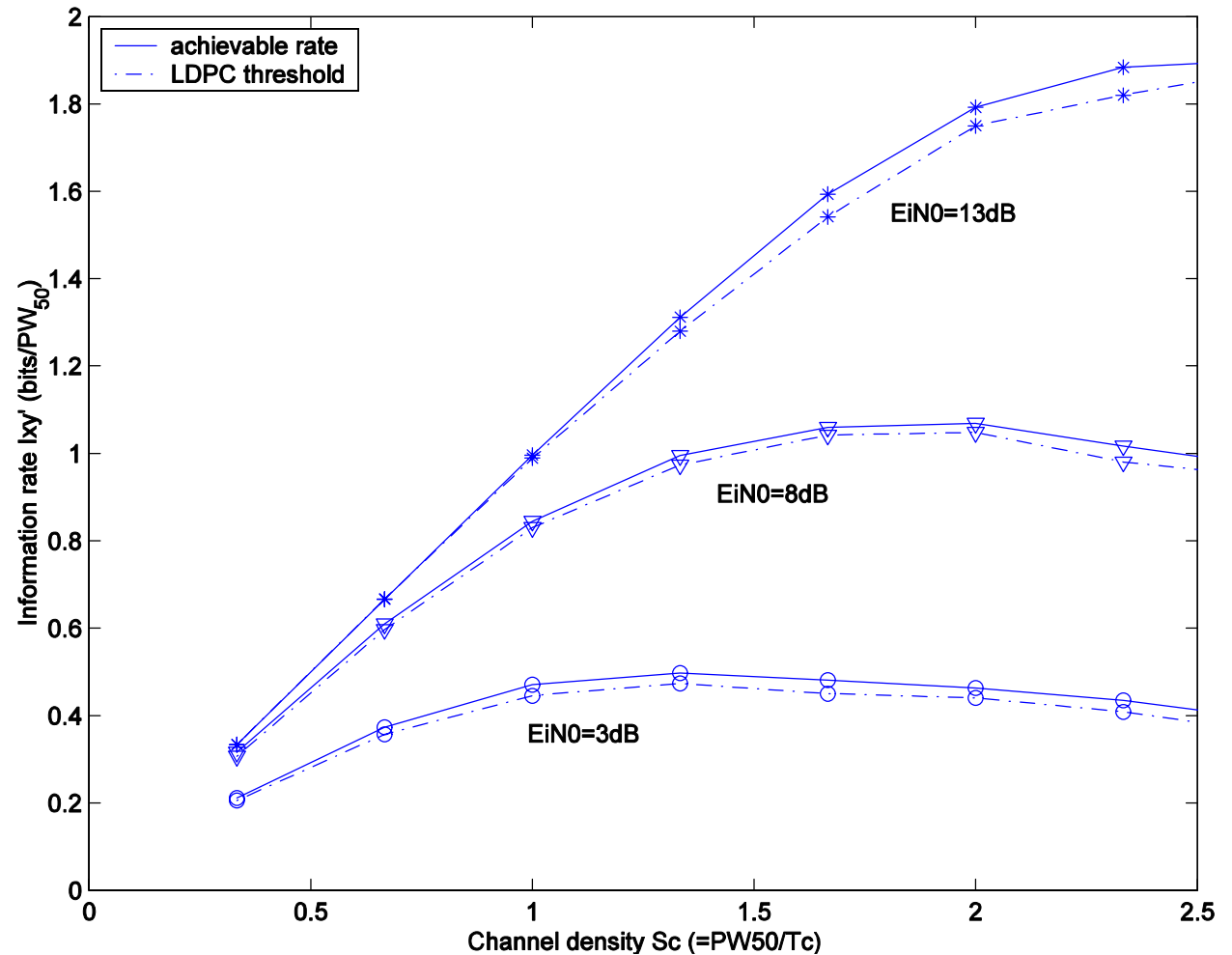
Results for LDPC Codes

Information rate $I(X;Y)$ for Lorentzian channel versus channel density - Shannon codes and LDPC codes.



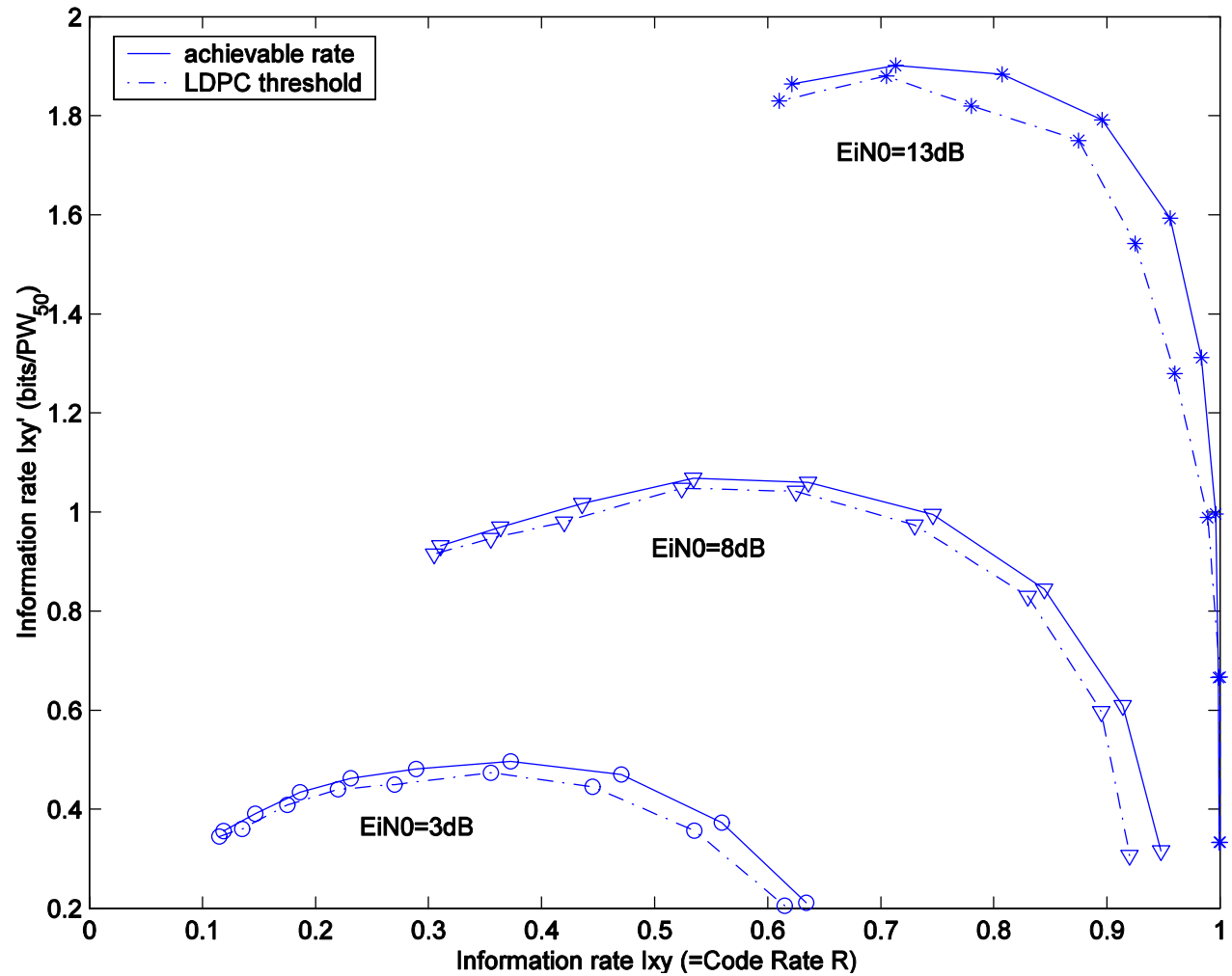
Results for LDPC Codes (cont'd)

Scaled Information rate $I'(X;Y)$ for Lorentzian channel versus channel density - Shannon codes and LDPC codes.



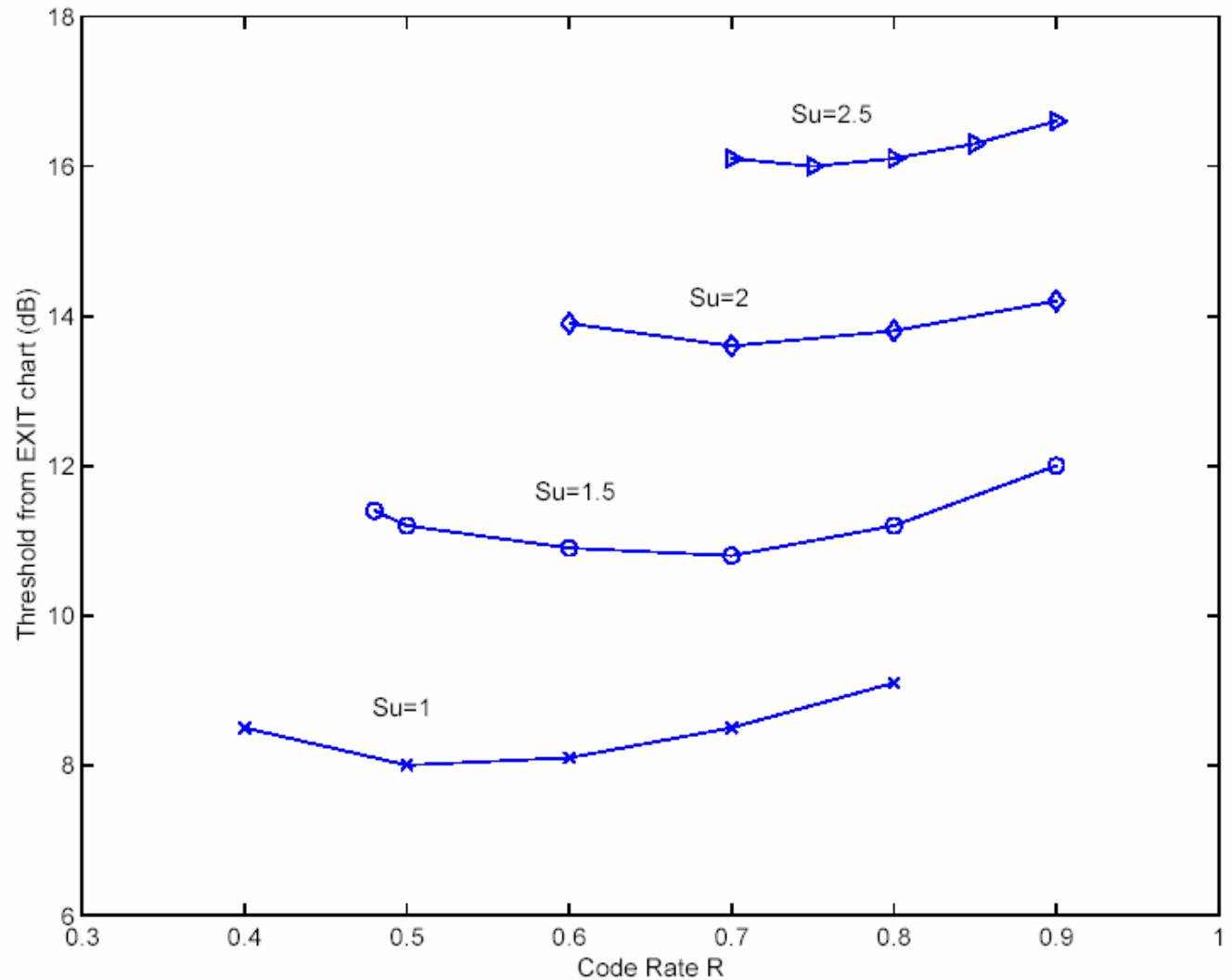
Results for LDPC Codes (cont'd)

Scaled Information rate $I'(X;Y)$ for Lorentzian channel versus code rate - Shannon codes and LDPC codes.



Results for LDPC Codes (cont'd)

Decoding threshold vs. code rate for various user densities



On the Optimal Bit Aspect Ratio

- ⌘ The information-theoretic areal density may be computed via

$$I_{areal} \text{ (bits/nm}^2\text{)} = I'_{xy} \text{ (bits/PW}_{50}\text{)} / [L_{50} \text{ (nm/PW}_{50}\text{)} \times TW^{-1} \text{ (tracks/nm)}]$$

where L_{50} is the length of PW_{50} in nm and TW is the track width.

- ⌘ It is well-known that the SNR along a track is proportional to the bit-length² under the Lorentzian model (Bergmans, Immink)
- ⌘ One may argue that at the optimal track density (which maximizes areal density), SNR will be proportional to the bit-width² as well (let bit-width = TW):

$$SNR = \alpha TW^2$$

Optimal Bit Aspect Ratio (cont'd)

- ⌘ Combining these two equations yields

$$I_{areal} = \sqrt{\alpha} \cdot I'_{xy} / (L_{50} \sqrt{SNR})$$

- ⌘ Since α and L_{50} are constants dependent on a specific hard disk drive, we define a normalized areal density measure

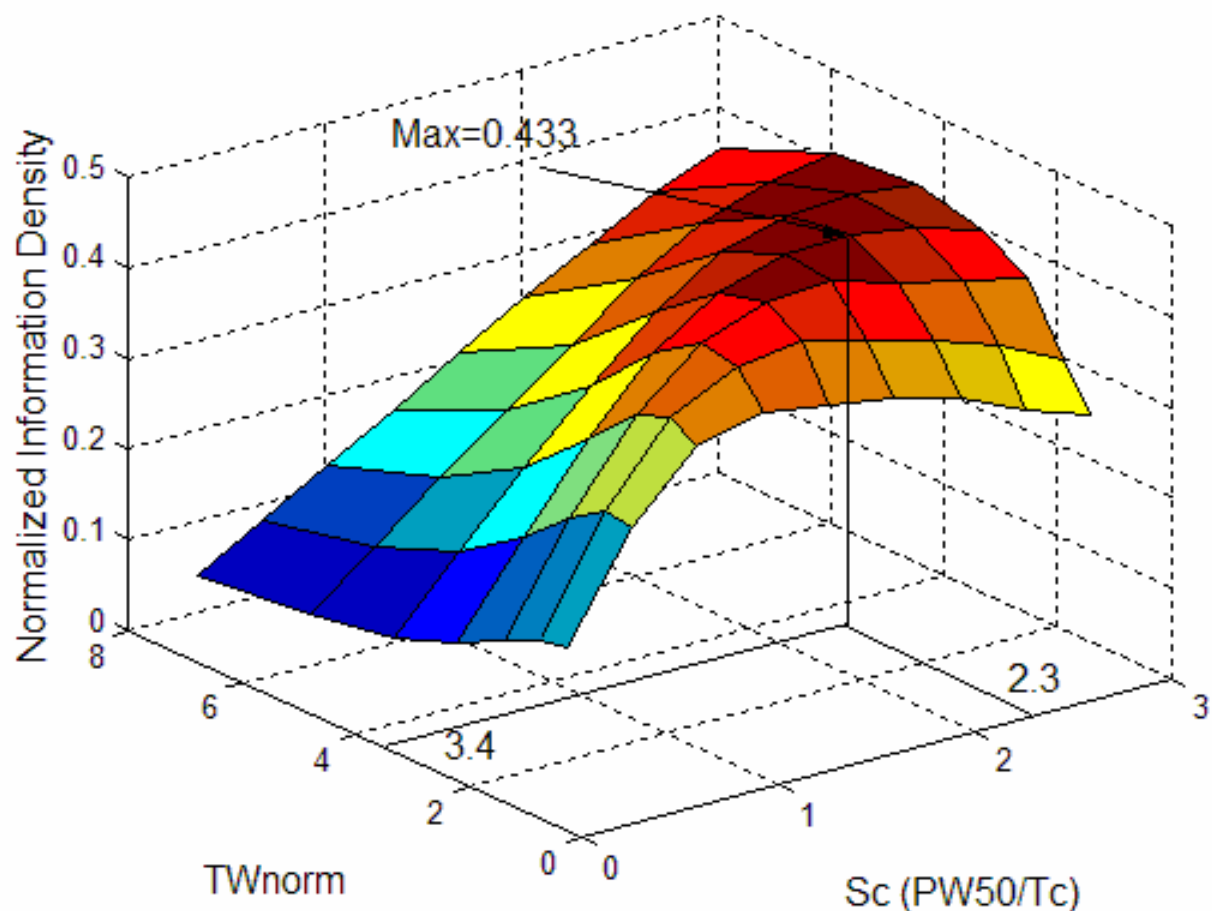
$$I_{areal,norm} = I_{areal} / (\sqrt{\alpha} / L_{50}) = I'_{xy} / \sqrt{SNR}$$

- ⌘ We may plot $I_{areal,norm}$ as a function of S_c (since I'_{xy} is a function of S_c) and the normalized track width $\sqrt{\alpha} TW$ (since \sqrt{SNR} in the previous equation may be replaced by $\sqrt{\alpha} TW$).

Optimal Bit Aspect Ratio (cont'd)

$I_{areal,norm}$ is maximized at $TW_{norm} = 3.4$ and $S_c = 2.3$.

We could convert $I_{areal,norm,max} = 0.433$ to a density in bits/in² by scaling this value by the factor $\sqrt{\alpha} / L_{50}$, if known.



Optimal Bit Aspect Ratio (cont'd)

⌘ Even in the absence of knowledge of a density measure in bits/in², this analysis yields the following operating values at the optimum:

⊠ SNR: $E_i/N_0 = 10.5$ dB

⊠ Code rate: $R = 0.62$

⊠ Channel density: $S_c = 2.35$

⊠ User density: $S_u = 1.45$

⌘ For comparison, today's (approximate) values:

⊠ SNR: $E_i/N_0 = 18$ dB

⊠ Code rate: $R = 0.95$

⊠ Channel density: $S_c = 3.0$

⊠ User density: $S_u = 2.85$

Conclusion

- ⌘ These results serve as a guide to choosing the optimal operating parameters (linear density, bit aspect ratio, code rate, etc.).
- ⌘ This work can be extended to include media noise and/or perpendicular recording.
- ⌘ It can also be extended to codes which do not have iid inputs (e.g., Markovian codes).
- ⌘ One of the implications is that work toward increased areal densities should target bit-width, not bit-length, leading to new challenges in track servo design.